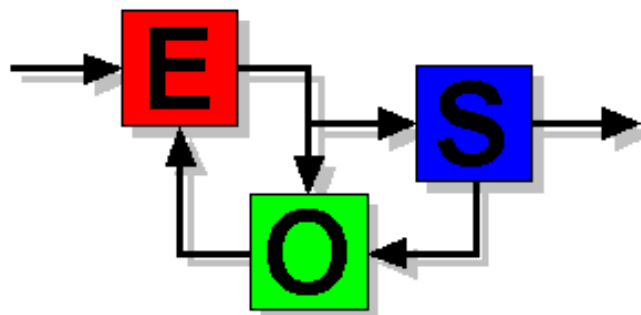


# Stationary Processes and Linear Systems

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# PART I: The general framework

## 1. Introduction

Time Series Analysis: Systematic approaches to extract information from time series, i.e., from observations ordered *in time* (no permutation invariance).

time series  $y_t, t = 1, \dots, T; y_t \in \mathbb{R}^n$

discrete, equidistant observations

- Data driven modeling
- Signal and feature extraction

Observations may be “noisy”.

Questions: Trends,

Hidden periodicities,

Dependence on time (dynamics)

Models: Stationary processes

Linear systems

## 2. The History of Time Series Analysis

### 2.1. The Early History (1772 - 1920)

- Late 18<sup>th</sup> century astronomy:
  - More accurate data from observation of the orbits of the planets
  - Kepler's laws are based on the two body problem—→ Are there deviations from the elliptic shape of the orbits (beside measurement noise)?  
hidden periodicities or trends

Question of secular changes (Laplace 1787; Jupiter, Saturn) Harmonic analysis:

- J. L. Lagrange (1736 - 1813), Oeuvres, Vol 6, 1772
- L. Euler (1707 - 1783)
- J. B. J. Fourier (1768 - 1831), Théorie analytique de la chaleur

- Method of least squares for fitting a line into a scatter plot: A.M. Legendre and C.F. Gauss: Early 19<sup>th</sup> century.
- Periodogramm: G.C. Stokes (1879), A. Schuster (1894) Detection of hidden periodicities:

$$I_T(\lambda) = \frac{1}{T} \left| \sum_{t=1}^T x_t e^{-i\lambda t} \right|^2$$

$T$  . . . sample size

Sunspot numbers, periodicity of earthquakes

- Empirical analysis of business cycles  
W.S. Jevons: Periodic fluctuations in economic time series (~ 1870, 1880)  
H. Moore “Economic Cycles: Their Law and Cause”  
1914  
W. Beveridge 1922: Wheat price index

## 2.2. The formation of modern time series analysis (1920-1970)

- Business cycles: Not exactly periodic:  
Stochastic models: AR and MA process e.g.

$$y_t = ay_{t-1} + \epsilon_t, \quad y_t = \epsilon_t + b\epsilon_{t-1}$$

$(\epsilon_t)$  white noise

G.U. Yule (1921, 1927) E. Slutsky (1927)

R. Frisch: Propagation and Impulse Problems in  
Dynamic Econometrics (1933)

- Theory of stationary processes:  
Concept: A. Ya Khinchin (1934)  
Spectral representation: A. N. Kolmogorov (1939,  
1941)  
Wold representation: H. Wold (1938)  
Factorization of spectra; linear least squares  
forecasting and filtering A.N. Kolmogorov (1939,  
1941)

Ergodic theory: G.D. Birkhoff (1931), A. Ya Khinchin (1932)

- Cowles Commission, Identifiability and ML estimation of (multivariate) ARX models.  
H.B. Mann and A. Wald (1943),  
T. Haavelmo (1944),  
T.C. Koopmann, H. Rubin and R.B. Leipnik (1950);  
Klein I model
- Spectral Estimation:  
Daniel (1946),  
R.B. Blackman and J. Tukey (1958),  
U. Grenander and M. Rosenblatt (1958),  
E. J. Hannan (1960)
- Asymptotic Theory for (mainly SISO) AR and ARMA estimation: Durbin, E.J. Hannan (1970), T.W. Anderson (1971)

### 2.3. The recent past (1970-1990)

- Box, G.E.P. and G.M. Jenkins (1970)  
Explicit instructions for SISO system identification:  
Differencing, Order determination, ML-estimation,  
validation
- Kalman: Structure theory for state space systems:  
Realization and parametrization. MIMO case
- Order estimation by information criteria such as AIC  
or BIC: Akaike, Hannan, Rissanen, Schwartz
- Asymptotic properties of ML-type estimation: E.J.  
Hannan (1973), W. Dunsmuir and E.J. Hannan  
(1976), P. Caines and L. Ljung (1979)
- Textbooks (late 80ies): Ljung, Caines, Hannan and  
Deistler, Söderström and Stoica

### 3. Areas of application

- Signal processing
- Control
- Econometrics: Macroeconometrics, finance, microeconometrics, marketing, logistics
- Medicine and biology



## PART II: Stationary Processes

### 4. Stationary processes in time domain

For us a stochastic process is a model for random phenomena evolving in time

$(\Omega, \mathcal{A}, \mathbb{P})$  probability space

$y_t : \Omega \rightarrow \mathbb{C}^n$  random variable

$(y_t | t \in T)$ , random process,  $T \subset \mathbb{R}$

in particular  $T = \mathbb{Z}$

Def . : A stochastic process  $(y_t)$  is called (weakly) *stationary* if:

(i)  $\mathbb{E}y_t^* y_t < \infty \quad t \in \mathbb{Z}$

(ii)  $\mathbb{E}y_t = m = \text{const} \quad t \in \mathbb{Z}$

(iii)  $\gamma(s) = \mathbb{E}y_{t+s} y_t^*$  does not depend on  $t$

Covariance function

$$\gamma : \mathbb{Z} \rightarrow \mathbb{C}^{n \times n} : \gamma(t) = \mathbb{E}y_t y_0^*$$

describes all linear dependence relations between the one dimensional random variables  $y_t^{(i)}, y_s^{(j)}$

$\gamma$  is a covariance function if and only if  $\gamma$  is nonnegative definite

$$y_t^{(i)} \in L_2 \text{ (over } (\Omega, \mathcal{A}, \mathbb{P}))$$

Note  $L_2$  with inner product

$$\langle x, y \rangle = \mathbb{E}x\bar{y}$$

is a Hilbert space

Def . : The *time domain*  $H \subset L_2$  of a stationary process  $y_t$  is the Hilbert space spanned by

$$\{y_t^{(i)} \mid t \in \mathbb{Z}, i = 1, \dots, n\}$$

## 5. Stationary processes in frequency domain

As a consequence of the “translation invariance” of the covariances we have:

**Theorem:** For every stationary process  $(y_t)$  there is a unique unitary operator  $U : H \rightarrow H$  such that  $y_t^{(i)} = U^t y_0^{(i)}$ ,  $i = 1, \dots, n$ ,  $t \in \mathbb{Z}$  holds.

From this we obtain

**Theorem:** (Spectral representation of stationary processes). For every stationary process  $(y_t)$  there exists a process  $(z(\lambda) | \lambda \in [-\pi, \pi])$  (called process with orthogonal increments)

satisfying

$$(i) \quad z(-\pi) = 0, \quad z(\pi) = x_0$$

$$(ii) \quad \lim_{\epsilon \downarrow 0} z(\lambda + \epsilon) = z(\lambda)$$

$$(iii) \quad \mathbb{E} z^*(\lambda) z(\lambda) < \infty$$

$$(iv) \quad \mathbb{E}\{(z(\lambda_4) - z(\lambda_3))(z(\lambda_2) - z(\lambda_1))^*\} = 0$$

for  $\lambda_1 < \lambda_2 \leq \lambda_3 < \lambda_4$

such that

$$y_t = \int e^{i\lambda t} dz(\lambda)$$

holds.

Thus, every stationary process is obtained as a limit of harmonic processes

$$y_t = \sum_{j=1}^h e^{i\lambda_j t} z(\lambda_j)$$

Second moments in frequency domain:

Spectral distribution function

$$F : [-\pi, \pi] \rightarrow \mathbb{C}^{n \times n} : F(\lambda) = \mathbb{E}z(\lambda)z(\lambda)^*$$

Spectral representation of covariance function

$$\gamma(t) = \int e^{i\lambda t} dF(\lambda)$$

$$\gamma \longleftrightarrow F$$

Spectral density (w. r. t. L-measure)

$$F(\lambda) = \int_{-\pi}^{\lambda} f(\omega) d\omega$$

exists e.g. if  $\sum \|\gamma(t)\|^2 < \infty$

$$\gamma(t) = \int e^{i\lambda t} f(\lambda) d\lambda$$

$$f(\lambda) = (2\pi)^{-1} \sum \gamma(t) e^{-i\lambda t}$$

$f$  (if it exists) contains the same information as  $\gamma$  but, often displayed in a more convenient form.

Peaks of  $f$  indicate dominating frequency bands.

## 6. Linear transformations of stationary processes

Let  $(x_t)$  be stationary; a linear transformation of  $(x_t)$  is given by  $y_t = \sum_{j=-\infty}^{\infty} k_j x_{t-j}$ ;  $k_j \in \mathbb{R}^{n \times m}$ ;

$$\sum \|k_j\| < \infty$$

then  $(x'_t, y'_t)'$  is jointly stationary.

$(k_j | t \in \mathbb{Z})$  weighting function

$$y_t = \int e^{i\lambda t} dz_y(\lambda) = \sum k_j \int e^{i\lambda(t-j)} dz_x(\lambda) =$$

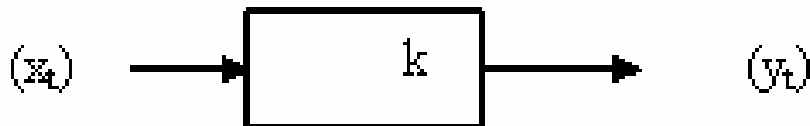
$$\int e^{i\lambda t} \left( \underbrace{\sum_{j=-\infty}^{\infty} k_j e^{-i\lambda j}}_{\text{transfer function } k(e^{-i\lambda})} \right) dz_x(\lambda)$$

transfer function  $k(e^{-i\lambda})$

$$k \longleftrightarrow (k_j)$$

The transfer function describes the linear transformation in frequency domain.

Linear system

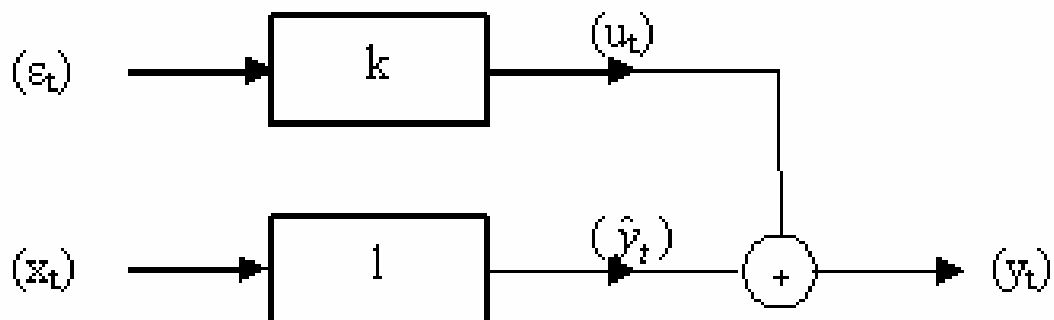


stable, time invariant

$(x_t)$  input

$(y_t)$  output

Linear system with noise



$(\epsilon_t)$  white noise, i.e.,  $\mathbb{E}\epsilon_t = 0$ ,  $\mathbb{E}\epsilon_s \epsilon_t' = \delta_{st} \cdot \Sigma$

$f_\epsilon = (2\pi)^{-1} \cdot \Sigma$

$$"dz_y(\lambda) = k(e^{-i\lambda})dz_x(\lambda)"$$

Transformation of second moments in frequency domain

$$f_y = k f_x k^*$$

$$f_{yx} = k f_x$$

## 7. The Wold decomposition

Let  $(x_t)$  be stationary

$H_x(t) = sp\{x_s^{(i)} | i = 1, \dots, n, s \leq t\} \subset H_x$  is called the past of  $(x_t)$

Def . : A stochastic process  $(x_t)$  is called

(linear) *regular* if  $\bigcap_t H_x(t) = \{0\}$

and (linear) *singular* if  $\bigcap_t H_x(t) = H_x$



Theorem: (Wold decomposition)

- (i) Every stationary process  $(x_t)$  can be uniquely decomposed as

$$x_t = y_t + z_t$$

where  $(y_t)$  is regular,  $(z_t)$  is singular,  $\mathbb{E}y_t z_s^* = 0$ ,  
 $y_t^{(i)} \in H_x(t)$ ,  $z_t^{(i)} \in H_x(t)$

- (ii) Every regular process  $(y_t)$  can be represented as

$$y_t = \sum_{j=0}^{\infty} k_j \epsilon_{t-j}, \quad \sum \|k_j\|^2 < \infty \quad (1)$$

where  $(\epsilon_t)$  is white noise satisfying  $\epsilon_t^{(i)} \in H_y(t)$ ,  
 $i = 1, \dots, n$

Thus, “practically every” stationary process is obtained as an output of a linear system whose input is the “simplest” random process, namely white noise.

$(\epsilon_s, s \leq t)$  constitutes an “orthonormal” basis for  $H_y(t)$ , (1) is an abstract Fourier series.

Linear least squares forecasting:

of  $y_{t+\tau}$  based on  $y_s, s \leq t$

project  $y_{t+\tau}^{(i)}$  on  $H_y(t)$ :

$$y_{t+\tau} = \underbrace{\sum_{j=h}^{\infty} k_j \epsilon_{t+\tau-j}}_{\text{forecast } \hat{y}_{t,\tau}} + \underbrace{\sum_{j=0}^{h-1} k_j \epsilon_{t+\tau-j}}_{\text{forecasting error}}$$

Spectral factorization:

The spectral density of  $(y_t)$  is given by

$$f_y = (2\pi)^{-1} \cdot \underbrace{\left( \sum_j k_j e^{-i\lambda j} \right)}_{k(e^{-i\lambda})} \cdot \Sigma \cdot k(e^{-i\lambda})^* \quad (2)$$

Question: Obtain  $k$  (and  $\Sigma$ ) from  $f_y$  (in 2).

## 8. Estimation I

Estimation of mean

$$\bar{y}_T = \frac{1}{T} \sum_{t=1}^T y_t$$

Estimation of covariances (ass.  $\mathbb{E}y_t = 0$ )

$$\hat{\gamma}(t) = \frac{1}{T} \sum_s y_{t+s} y_s^*$$

Estimation of spectra

Periodogram

$$I_T(\lambda) = \sum \hat{\gamma}(t) e^{-i\lambda t}$$

is not consistent; smoothed periodogram

## 9. Rational spectra, ARMA and state-space systems

### ARMA system

$$\underbrace{\sum_{j=0}^p a_j y_{t-j}}_{a(z)y_t} = \sum_{j=0}^p b_j \epsilon_{t-j} = b(z)\epsilon_t$$

$z$ : backward shift as well as complex variable

### Stability condition

$$\det a(z) \neq 0 \quad |z| \leq 1$$

### Miniphase condition

$$\det b(z) \neq 0 \quad |z| < 1$$

### Normalization

$$a_0 = b_0$$

### Steady state solution

$$y_t = \underbrace{a^{-1}(z)b(z)}_{\text{Transfer function } k(z)} \epsilon_t$$

This transfer function corresponds to the Wold decomposition.

### State space system

$$x_{t+1} = Ax_t + B\epsilon_t$$

$$y_t = Cx_t + \epsilon_t$$

$x_t$ : state (n-dimensional)

Stability condition

$$|\lambda_{max}(A)| < 1$$

Miniphase condition

$$|\lambda_{max}(A - BC)| \leq 1$$

Steady state solution

$$y_t = (C(Iz^{-1} - A)^{-1}B + I)\epsilon_t$$

again corresponds to the Wold decomposition

Theorem :

- (i) Every rational and a.e. nonsingular spectral density matrix may be uniquely factorized (as in (2)), where  $k(z)$  is rational (in  $z \in \mathbb{C}$ ), analytic within a circle containing the closed unit disk,  $\det k(z) \neq 0$ ,  $|z| < 1$  and  $\Sigma > 0$ .
- (ii) For every rational transfer function  $k$  satisfying the above mentioned properties there is a stable and miniphase ARMA system with  $a_0 = b_0$  and conversely every such ARMA system has a transfer function with the properties mentioned in (i).
- (iii) A completely analogous statement holds for state space systems

## PART III: Identification of linear systems

### 10. Problem statement

Data driven modeling: Find a good model from (noisy) data

One has to specify:

- The model class, i.e. the class of all a priori feasible candidate systems to be fitted to the data. Here the model class is the set of all stable and miniphase ARMA or state space systems (for given  $s$ )
- The class of feasible data
- An identification procedure, which is a set of rules - in the fully automatized case a function - attaching to every feasible data string  $y_t, t = 1, \dots, T$  a system from the model class.

The theory of identification is mainly concerned with the development and evaluation of identification algorithms.

### Steps in actual identification

- Data generation and preprocessing of data (e.g. removing outliers)
- Description of the model class using the prior knowledge available
- Identifying the model
- Model validation

### In identification in general the following parameters have to be determined from data

- Integer-valued parameters such as the state dimension of a minimal state space system; this defines a subclass, namely the class of all systems of order  $n$
- Real-valued parameters, such as the entries in (A,B,C)

### Semi-nonparametric estimation problem



3 modules of the problem:

- Structure theory: Idealized identification, we commence from the population second moments of the observations or from the (“true”) transfer functions rather than from data
- Estimation of real-valued parameters for given integer-valued parameters
- Model selection: Estimation of integer-valued parameters

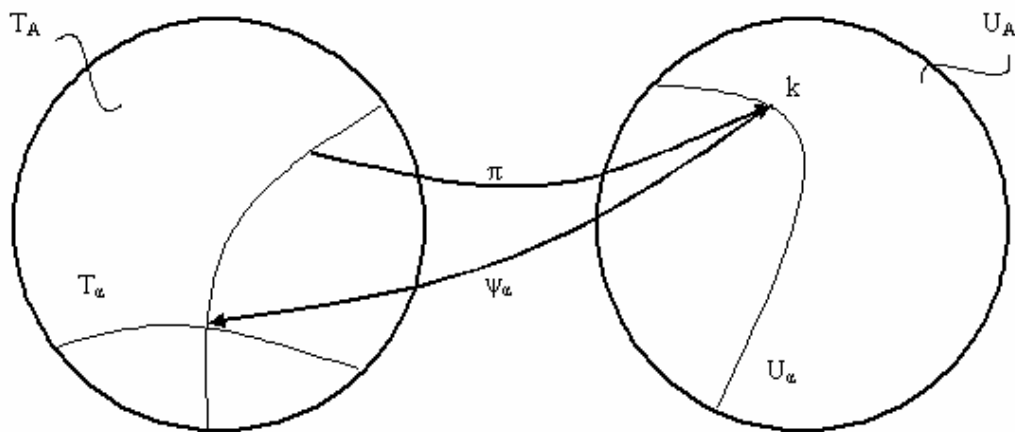
# 11. Structure Theory

Note: (2) defines a one-to-one relation between  $f$  and  $k$ ,  $\Sigma$  under our assumptions.

We restrict ourselves to state space systems:

Let  $U_A$  denote the set of all rational  $s \times s$  transfer functions  $k(z)$  satisfying our assumptions and let  $T_A$  denote the set of all state space systems  $(A, B, C)$  (for fixed  $s$  but variable  $n$ ) satisfying our assumptions; finally let the mapping  $\pi : T_A \rightarrow U_A$  be defined by

$$\pi(A, B, C) = C(Iz^{-1} - A)^{-1}B + I$$



- $\pi$  is not injective (no identifiability)  
 $\pi^{-1}(k)$  the class of observationally equivalent systems
- There exists no continuous selection of representatives from the equivalence classes

“illposedness”

Identifiability and continuity of the parametrization are desirable:  $U_A$  and  $T_A$  are broken into bits  $U_\alpha$  and  $T_\alpha$  respectively s.t.  $\pi/T_\alpha$  is injective and surjective and its inverse, the parametrization

$$\psi_\alpha : U_\alpha \rightarrow T_\alpha \subset \mathbb{R}^{d_\alpha}$$

is continuous

Free parameters  $\mathbb{R}^{d_\alpha} \ni \tau_\alpha \leftrightarrow (A, B, C)$

for given  $T_\alpha$ .

## 12. Estimation of real-valued parameters

(Gaussian) Likelihood function ( $-2T^{-1} \times \log$ )

$$L_T(\tau_\alpha, \Sigma) = T^{-1} \log \det \Gamma_T(\tau_\alpha, \Sigma) \\ + T^{-1} y'(T) \Gamma_T^{-1}(\tau_\alpha, \Sigma) y(T)$$

where

$$y(T) = (y'_1, \dots, y'_T)' \text{ (stacked sample)}$$

$$\underbrace{\Gamma_T(\tau_\alpha, \Sigma)}_{T \cdot s \times T \cdot s} = \left( \int e^{-i\lambda(r-t)} f_y(\lambda; \tau_\alpha, \Sigma) d\lambda \right)_{r,t=1,\dots,T}$$

ML estimators:

$$(\hat{\tau}_{\alpha,T}, \hat{\Sigma}_T) = \arg \min_{\tau_\alpha \in T_\alpha, \Sigma \in \underline{\Sigma}} L_T(\tau_\alpha, \Sigma)$$

Coordinate free MLE:  $\hat{k}_T$

Asymptotic properties:

- Consistency:

$$\begin{aligned}\hat{k}_T &\rightarrow k_0 \\ \hat{\Sigma}_T &\rightarrow \Sigma_0\end{aligned}$$

- Asymptotic normality

$$\sqrt{T}(\hat{\tau}_{\alpha,T} - \tau_{\alpha,0}) \rightarrow N(0, V)$$

## 13. Model Selection

Example: Estimation of  $n$ , analogous for  $\alpha$

Information criteria: Tradeoff between fit and complexity

$$I_T(n) = \underbrace{\log \det \hat{\Sigma}_T(n)}_{\text{Measure of fit}} + \underbrace{(2ns)}_{\text{dimension, measures complexity}} \frac{c(T)}{T}$$

$$\hat{n}_T = \arg \min I_n$$

$$\text{AIC criterion } c(T) = 2$$

$$\text{BIC criterion } c(T) = \log(T)$$

BIC is consistent, AIC not

Post model selection properties