Nonlinear Filter Design for Pose and IMU Bias Estimation

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Introduction

- Context:
 - High quality pose estimation \rightarrow orientation and position.
 - Data fusion between different sources... maximizes virtues and minimizes drawbacks of each sensor!
 - Extreme case of fast dynamics, e.g. MUAVs.
- IMU (Inertial Measurement Unit):
 - High frequency: 50 to 1k [Hz].
 - Incremental measurements.
 - angular rate gyroscopes, accelerometers.
 - Drawbacks: measurement high frequency noise and additive offset.
 - Recover pose by IMU integration: drifts quickly.
- Pose measurements:
 - Advantages: no drift.
 - Low frequency: generally from 1 up to 25 [Hz].

Nonlinear observer design

Orientation and
gyro bias dynamicsPosition and
accelerometer bias dynamics $\begin{cases} \frac{d}{dt}R = RS(\omega - \omega_b), \\ \frac{d}{dt}\omega_b = 0. \end{cases}$ (1) $\begin{pmatrix} \frac{d}{dt}\mu = v, \\ \frac{d}{dt}v = R(a - a_b) + g_{\mathcal{I}}, \\ \frac{d}{dt}a_b = 0. \end{cases}$

Measurements

$$y = (\omega, a, R, p).$$
 (3)



Error definition $\widetilde{R} \triangleq R\widehat{R}^T, \ \widetilde{\omega}_b \triangleq \omega_b - \widehat{\omega}_b,$ $\widetilde{p} \triangleq p - \widehat{p}, \ \widetilde{v} \triangleq v - \widehat{v}, \ \widetilde{a}_b \triangleq a_b - \widehat{a}_b.$

• The innovation terms should make $(\tilde{R}, \tilde{\omega}_b, \tilde{p}, \tilde{v}, \tilde{a}_b) = (I_3, 0, 0, 0, 0)$ an asymptotically stable equilibrium point of:

Error dynamics

$$\begin{cases} \frac{d}{dt}\widetilde{R} = \widetilde{R}S(-\widehat{R}\widetilde{\omega}_{b} - \widehat{R}\alpha_{R}), \\ \frac{d}{dt}\widetilde{\omega}_{b} = -\alpha_{\omega}, \\ \begin{cases} \frac{d}{dt}\widetilde{p} = \widetilde{v} - \alpha_{p}, \\ \frac{d}{dt}\widetilde{v} = -R\widetilde{a}_{b} - \alpha_{v}, \\ \frac{d}{dt}\widetilde{a}_{b} = -\alpha_{a}. \end{cases}$$
(6)

Orientation and gyro bias estimation

- Orientation and gyro bias dynamics is **independent** from position and accelerometer bias.
- Several solutions with semi-global stability...
- Passive complementary filter on **SO**(3) (Mahony, Hamel & Pflimlin 2008).

Lemma

Let

$$\alpha_{R} = k_{1} \widehat{R}^{T} \operatorname{vex}(\mathbf{P}_{\mathbf{a}}(\widetilde{R})), \ \alpha_{\omega} = -k_{2} \widehat{R}^{T} \operatorname{vex}(\mathbf{P}_{\mathbf{a}}(\widetilde{R})),$$

with $k_1, k_2 > 0$. Then, concerning the dynamics (6): 1) All solutions converge to $\mathbb{E}_s \cup \mathbb{E}_u$ with $\mathbb{E}_s = (I_3, 0)$, and $\mathbb{E}_u = \left\{ (\tilde{R}, \tilde{\omega}_b) \mid \operatorname{tr}(\tilde{R}) = -1 \right\}.$ 2) $(\tilde{R}, \tilde{\omega}_b) = (I_3, 0)$ is a locally exponentially stable equilibrium.

Nonlinear observer design

Position and accelerometer bias estimation

- First: R and $\omega_{\mathcal{B}}$ are available.
- Position and acceleration estimation:

Theorem

Let

$$\alpha_{p} = k_{3}\widetilde{p}, \ \alpha_{v} = k_{4}\widetilde{p}, \ \alpha_{a} = -k_{5}(I_{3} + \frac{1}{k_{3}}S(\omega_{\mathcal{B}}))R^{T}\widetilde{p}$$

with $k_3, k_4, k_5 > 0$ such that $k_5 < k_3k_4$. Then, $(\tilde{p}, \tilde{v}, \tilde{a}_b) = (0, 0, 0)$ is a globally exponentially stable equilibrium point of the position estimation error dynamics (7).

- This theorem yields a globally asymptotically stable estimator.
- Stability is achieved for any angular velocity ω_B.
- The initial assumption can be relaxed.

Nonlinear observer design

Full pose observer

Corollary



$$\begin{cases} \frac{d}{dt}\widehat{R} &= \widehat{R}S(\omega - \widehat{\omega}_b + \alpha_R), \\ \frac{d}{dt}\widehat{\omega}_b &= \alpha_\omega, \\ \begin{cases} \frac{d}{dt}\widehat{p} &= \widehat{v} + \alpha_p, \\ \frac{d}{dt}\widehat{v} &= g_{\mathcal{I}} + \widehat{R}(a - \widehat{a}_b) + \alpha_v, \\ \frac{d}{dt}\widehat{a}_b &= \alpha_a, \end{cases}$$

with $\alpha_R = k_1 \widehat{R}^T \operatorname{vex}(\mathbf{P}_{\mathbf{a}}(\widetilde{R})), \ \alpha_\omega = -k_2 \widehat{R}^T \operatorname{vex}(\mathbf{P}_{\mathbf{a}}(\widetilde{R})),$

$$\alpha_p = k_3 \tilde{p}, \ \alpha_v = k_4 \tilde{p}, \ \alpha_a = -k_5 (l_3 + \frac{1}{k_3} S(\omega - \widehat{\omega}_b)) \tilde{R}^T \tilde{p}.$$

Assume that $k_1, \dots, k_5 > 0$ and $k_5 < k_3k_4$. Then, 1) The origin $(\tilde{R}, \tilde{\omega}_b, \tilde{p}, \tilde{v}, \tilde{a}_b) = (l_3, 0, 0, 0, 0)$ is a locally exponentially stable equilibrium.

2) If \tilde{R} converges asymptotically to I_3 , then $(\tilde{\omega}_b, \tilde{p}, \tilde{v}, \tilde{a}_b)$ converges asymptotically to zero.

Tuning the innovation gains

- How to tune the innovation gains?
- Rationale: analyze the error dynamics and each gain's effect.
- Using this procedure, one can define 5 settling times $\tau_i > 0$, Gain tuning for the nonlinear observers: $k_1 = 3 \frac{\tau_1 + \tau_2}{\tau_1 \tau_2}, \quad k_2 = 9 \frac{1}{\tau_1 \tau_2},$ $k_3 = 3 \frac{\tau_3 \tau_4 + \tau_3 \tau_5 + \tau_4 \tau_5}{\tau_3 \tau_4 \tau_5}, \quad k_4 = 9 \frac{\tau_3 + \tau_4 + \tau_5}{\tau_3 \tau_4 \tau_5}, \quad k_5 = \frac{27}{\tau_3 \tau_4 \tau_5}.$
- This definition of k_3 , k_4 , and k_5 satisfies $k_5 < k_3 k_4$.

Results

Enhancing visual pose estimation using data fusion - Visual estimation at 40 [Hz].

Pose estimation using (Benhimane & Malis 2007). Visual update at 40 [Hz].



Results

Enhancing visual pose estimation using data fusion - Visual-IMU fusion at 200 [Hz].

The same visual pose estimation using the proposed filter. IMU at 200 [Hz], visual update at 10 [Hz].



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Conclusion

- Design of **nonlinear observers** to estimate pose with online calibration of IMU bias.
- Semi-global stability is achieved.
- Gain tuning method based on errors settling times.

Future work

- Evaluation of **time varying innovation gains**, and the use of gain matrices, also relating with **measurement and estimate uncertainties**.
- Extension for **coordinate system parameter estimation**, e.g. camera-to-IMU orientation and translation.

Bibliography

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